

Active Power Factor Correction Using Switching Regulators

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Abstract - The most of the equipments being used in IT industries, commercial and residential applications are internally working on the DC voltage. So either internal to the equipment or externally they need to undergo conversion from AC to DC for their functioning and operation. Depending on the requirement it consists of Single or Three Phase controlled or uncontrolled rectifier. The simple rectification technique leads to degradation in the quality of the current drawn in terms of Harmonics, wave form Distortion, Power Factor (PF) and affect the power distribution system. The increasing demand of these devices, the line current harmonics and PF pose a major problem by degrading the power quality of the system thus affecting the performance of the other devices also. Hence there is a need to enhance quality of power drawn by these equipments.

This paper proposes the front end rectification technique for three phase supply using switching mode active control to enhance these quality factors. The technique is simulated in the MATLAB Simulink environment and results are shown. It also focuses the approach of Pre-regulator integrated circuits towards the power factor correction.

Keywords—Active Power Factor Correction (APFC), Current Loop, Voltage Loop, Harmonic Current.

I. INTRODUCTION

The power supplies in most of the equipment are to convert mains power to the desired level of power that is needed for their proper functioning. This conversion leads to rectification and filtering of input mains power. Then it is scaled down using DC-DC converters as shown in Fig.1.

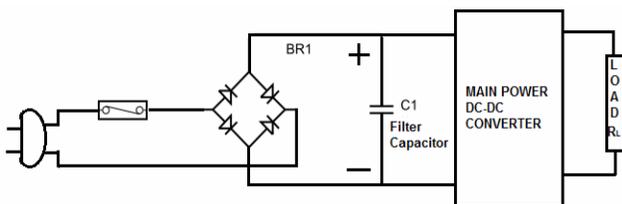


Figure 1: Capacitive line filter

The filter used in the supply draws current only when the line voltage exceeds the filter capacitor voltage and the filter capacitor is charged to near the peak level of the input line voltage as shown in figure 2.

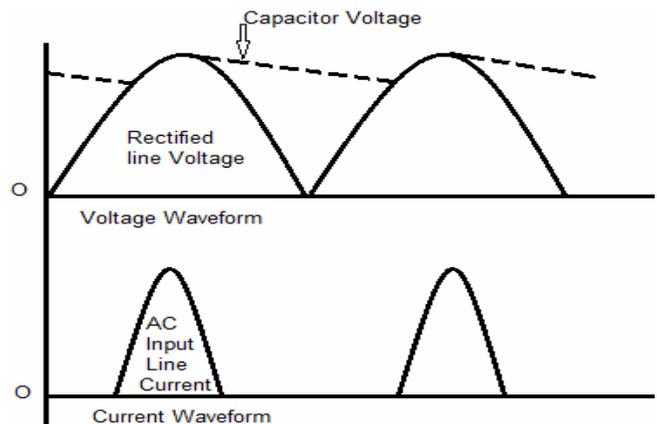


Figure 2: Input Current with Capacitor Filter

As shown in the Fig.2, the current flows for the short duration where the capacitor voltage reaches to peak of input voltage. Due to this short pulses the RMS value of current increases and average value decreases, which adds line losses and power quality issues in the system.

Power factor correction may lead to have benefits like; reduction in distribution cost, more power can be drawn at the same current rating from the line. Presently it is mandatory requirement in many countries for the equipment to comply the standards led by the government.

II. HOW DOES PFC WORKS?

The ideal requirement of power factor correction is making nature of input current waveform same as that of the input voltage. Figure 3 shows the line voltage and line current with and without power factor correction. Major two power factor correction techniques are Active PFC and Passive PFC.

A passive PFC uses a filter at the AC input to correct poor power factor. The passive PFC circuitry uses only passive components-an inductor and capacitors. A passive PFC rarely achieves low Total Harmonic Distortion (THD). Also, because the circuit operates at the low line power frequency of 50Hz or 60Hz, the passive elements are normally bulky and heavy [1].

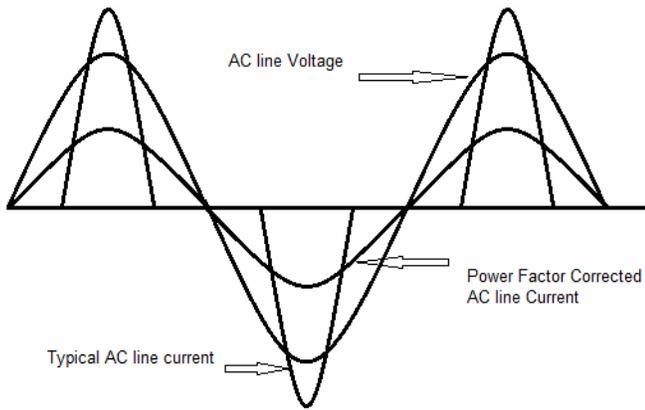


Figure3: Effect of PFC on line current.

Active PFC offers better THD and is significantly smaller and lighter than a passive PFC circuit. It uses switching regulator technology to draw current from the power line proportional to input voltage [2]. It consists of feedback loops to control input current and to regulate output voltage.

III. SWITCHING REGULATOR TOPOLOGIES

Buck, boost, Buck-boost and other converter topologies are used in active PFC circuits.[3], since the only requirement is to close a fast current loop around the input current and make it look like input voltage.

A) Buck Converter:

In this type of topology, the DC output voltage which appears across the load is a fraction of the input voltage and this fraction turns out to be equal to the duty cycle. So we can write:

$$V_{out} = V_{in} \times D$$

Where, D is duty cycle

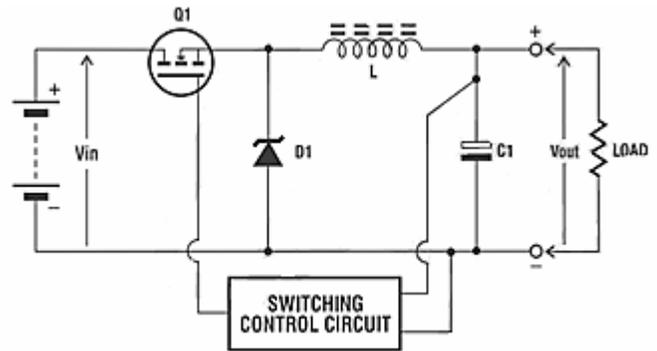


Figure 4: Buck Converter.

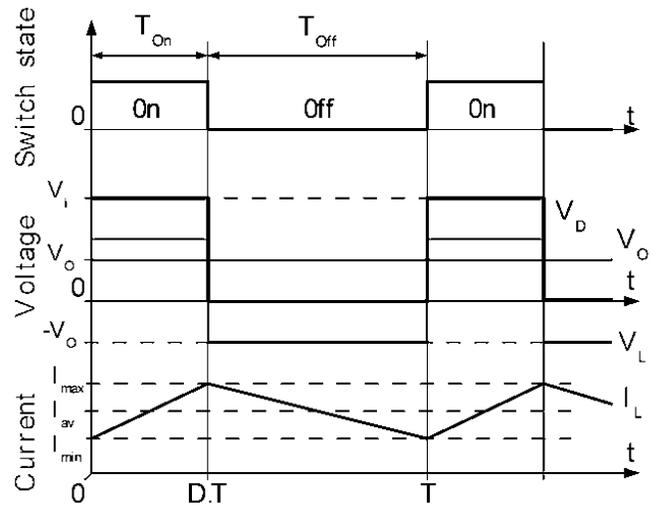


Figure 5: Waveforms of Current and Voltage in Buck Converter operating in continuous conduction mode

So by varying the switching duty cycle, the buck converter's output voltage can be varied as a fraction of the input voltage.

B) Boost Converter:

The operation of Boost converter [4] consists of using Q1 as high switching speed. When Q1 is switched on, current flows from the input source through L and Q1, and energy is stored in the inductor's magnetic field. There is no current

through D1, and the load current is supplied by the charge in C1. Then when Q1 is turned off, L opposes any drop in current by immediately reversing its EMF so that the inductor voltage adds to (i.e. boosts) the source voltage, and current due to this boosted voltage now flows from the source through L, D1 and the load, recharging C1 as well.

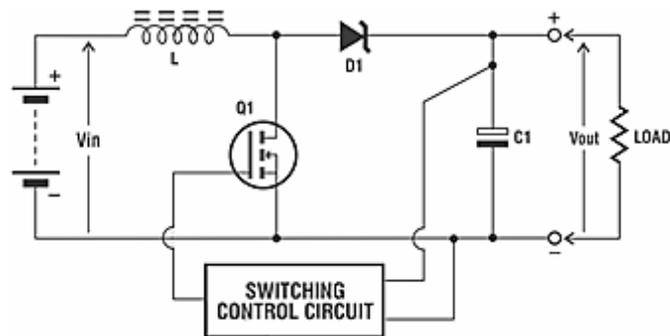


Figure 6: Boost Converter

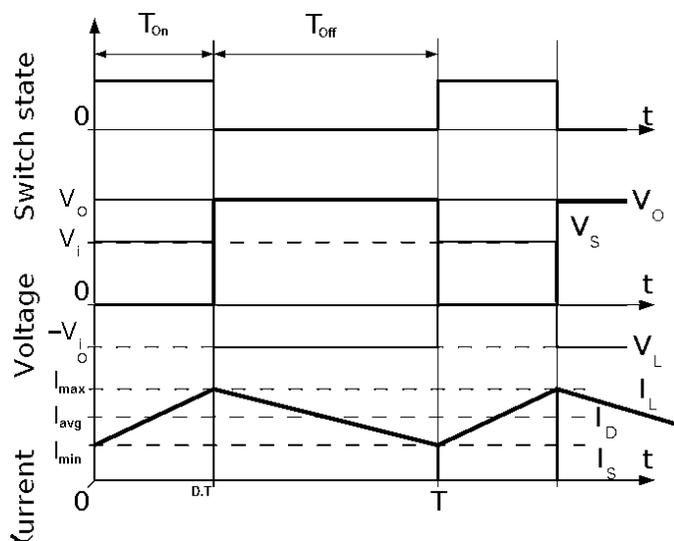


Figure 7: Waveforms of Current and Voltage in Boost Converter operating in continuous conduction mode

The output voltage is therefore higher than the input voltage and it turns out that the voltage step up ratio is equal to

$$V_{out}/V_{in} = 1/(1-D) \text{ or } V_{out}/V_{in} = T/T_{off}$$

Where, 1-D is actually the proportion of switching cycle that Q1 is off, rather than on.

C) Buck-boost Converter

This converter allows the voltage to be stepped either up or down [5], depending on the duty cycle. With this type of configuration the ratio between the output and input voltage turns out to be,

$$V_{out}/V_{in} = -D/(1-D) \text{ or } V_{out}/V_{in} = -T_{on}/T_{off}$$

So, the buck-boost converter steps the voltage down when the duty cycle is less than 50% ($T_{on} < T_{off}$) and steps up when duty cycle is more than 50% ($T_{on} > T_{off}$)

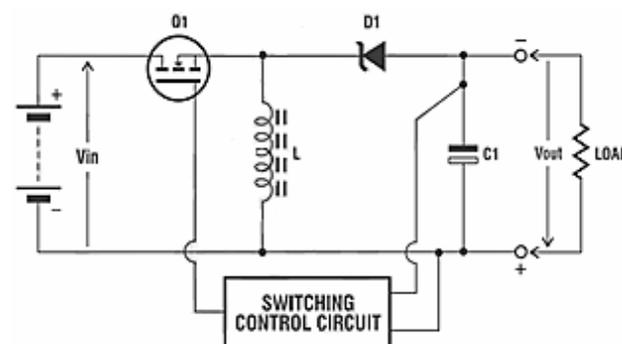


Figure 8: Buck-Boost Converter

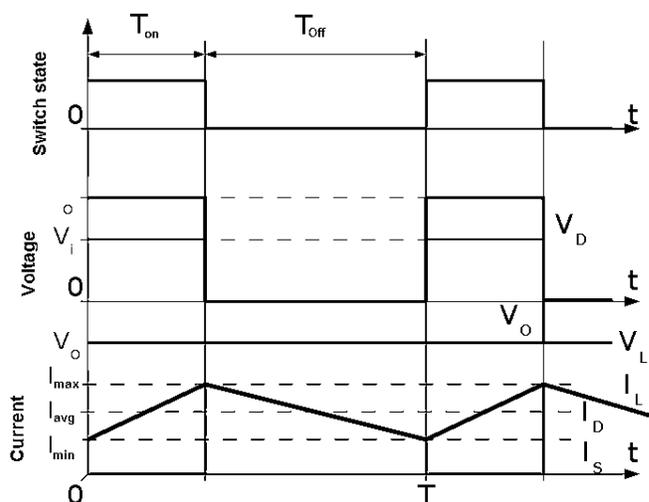


Figure 9: Waveforms of Current and Voltage in Buck-Boost Converter operating in continuous conduction mode

The most popular topology seems to be the boost converter [7], since the input current flows through an inductor and is relative smooth and easy to control and also easy to measure with a resistor in the power return. Another popular

approach, especially at lower power levels, is the buck-boost converter. This approach allows isolation and output voltage amplitude scaling as part of the power factor correction process.

IV. PREREGULATOR INTEGRATED CIRCUIT

Because of the popularity of power factor correction, a number of companies now manufacture power factor correction ICs. These include: Cherry Semiconductor Corp., IXYS Corp., Linear Technology Corp., Micro Linear Corp., Motorola, Inc. (Tempe, AZ), Unitrode Corp., etc.

The Unitrode UC3854 family [6] is typical and is used for the examples in this paper. This IC uses boost topology. Figure 7 is from the Unitrode data sheet and shows the UC3854 connected in a typical 250-watt power factor corrected pre-regulator.

A) Current Control Loop

The current control loop contains error amplifier which is used for comparison of sensed current (pin 4) and reference current which comes from the multiplication of output error voltage and absolute value of instantaneous line voltage. The reference takes the form of a current that is injected into the non-inverting input of the current error amplifier (pin 5).

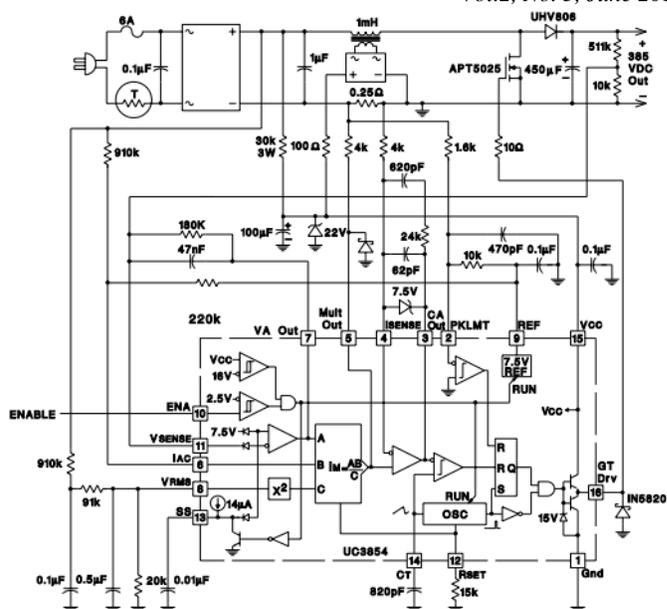


Figure 10: Schematic of 250 watt pre-regulator from Unitrode data sheet [6].

PWM pulses are then created by comparing output of error amplifier with fixed amplitude ramp signal (pin 14). These pulses are used to drive the power transistors (pin 16).

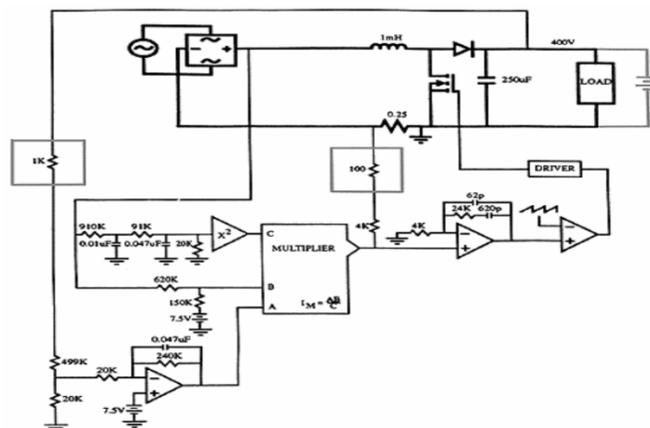


Figure 11: Simplified schematic of pre-regulator showing signal paths and loops [6].

B) Voltage Control Loop

The voltage loop is used to compensate changes in line and load voltages. The DC output is sensed (pin 11) and compared with fixed DC voltage in the error amplifier. The output of error amplifier (pin 7) is then converted into current reference waveform by multiplying it with a waveform template (pin 5), which represents the desired current wave shape. This multiplied output drives the current control loop. The DC output of squarer which is proportional to the square of V_m is used to pre-scale the error amplifier output. This makes open loop gain independent of line voltage.

In the Active filter operation, the nonlinear load shown in Fig. 12 is again a three phase filter [8]. It can be seen that the mains current gets asine waveform, being only slightly distorted in the instants of the diode switching. This is due to the high di/dt occurring in these points, which is impossible to compensate unless with a very high DC voltage or a very low AC inductance. There should be a compromise between the active filter dimensioning and the characteristics of the load current to be compensated.

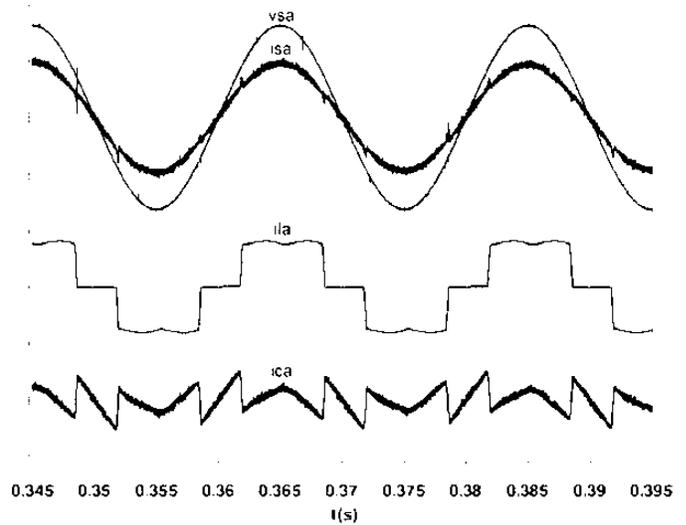


Figure 12: Steady-state compensation of a filter. Traces from top to bottom: source voltage, V_{sa} ; source current, I_{sa} ; load current, I_{la} ; filter current, I_{ca} [8].

V. SIMULATION MODEL

The APFC control scheme is tested under the MATLAB Simulink environment. The major components of the model are Voltage control loop, Current control loop, PWM generator, steering control [10] as shown in Fig. 13.

The model is developed considering the three phase load. It is directly connected to the line without interference to the existing load connection. Three phase IGBT based active rectifier is used as the power topology. The conduction of the IGBT is controlled based on the instantaneous correction depending on the current drawn at that instance by the connected load. the topology is tested for the lagging PF load conditions in the preliminary stage.

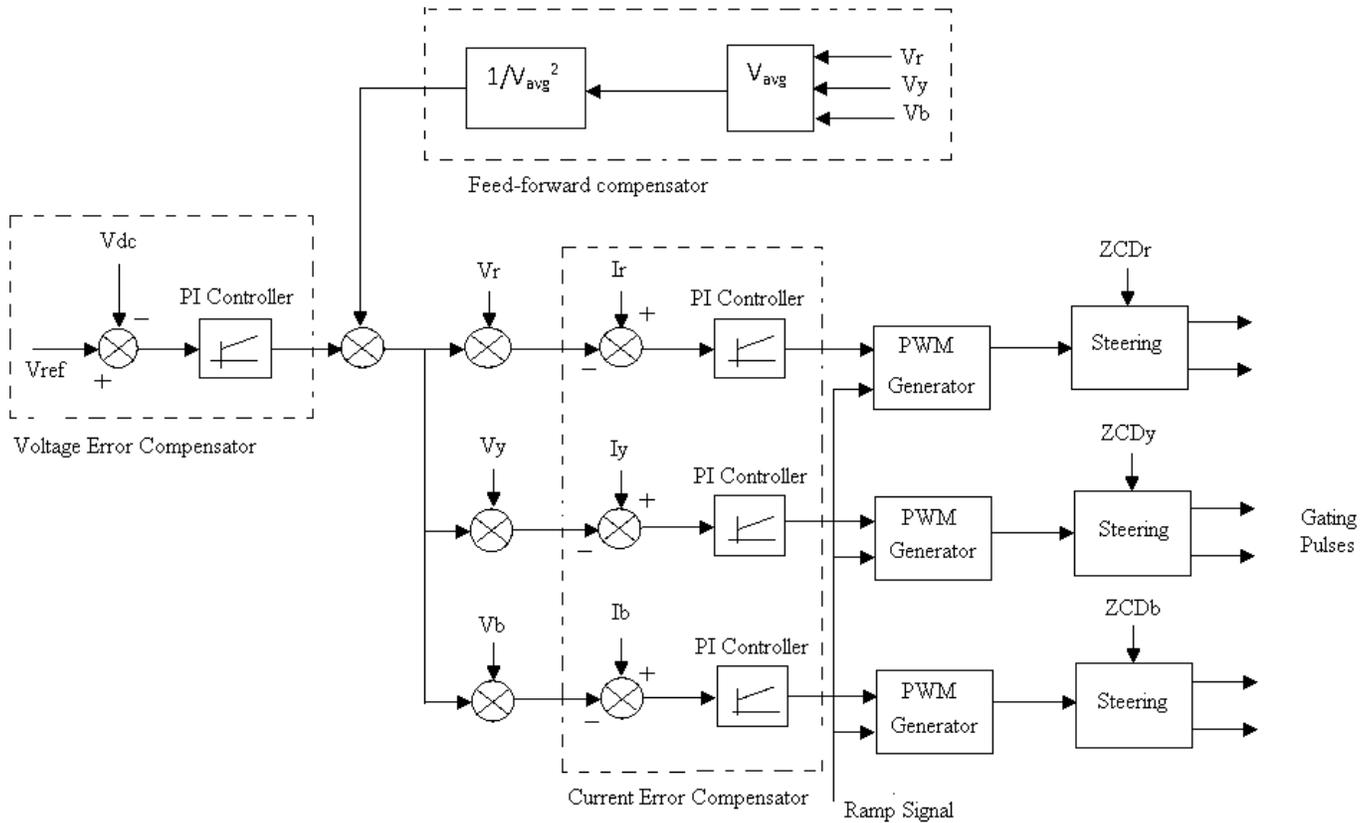


Figure 13: Simplified diagram of Control loops used in Preregulator circuit.

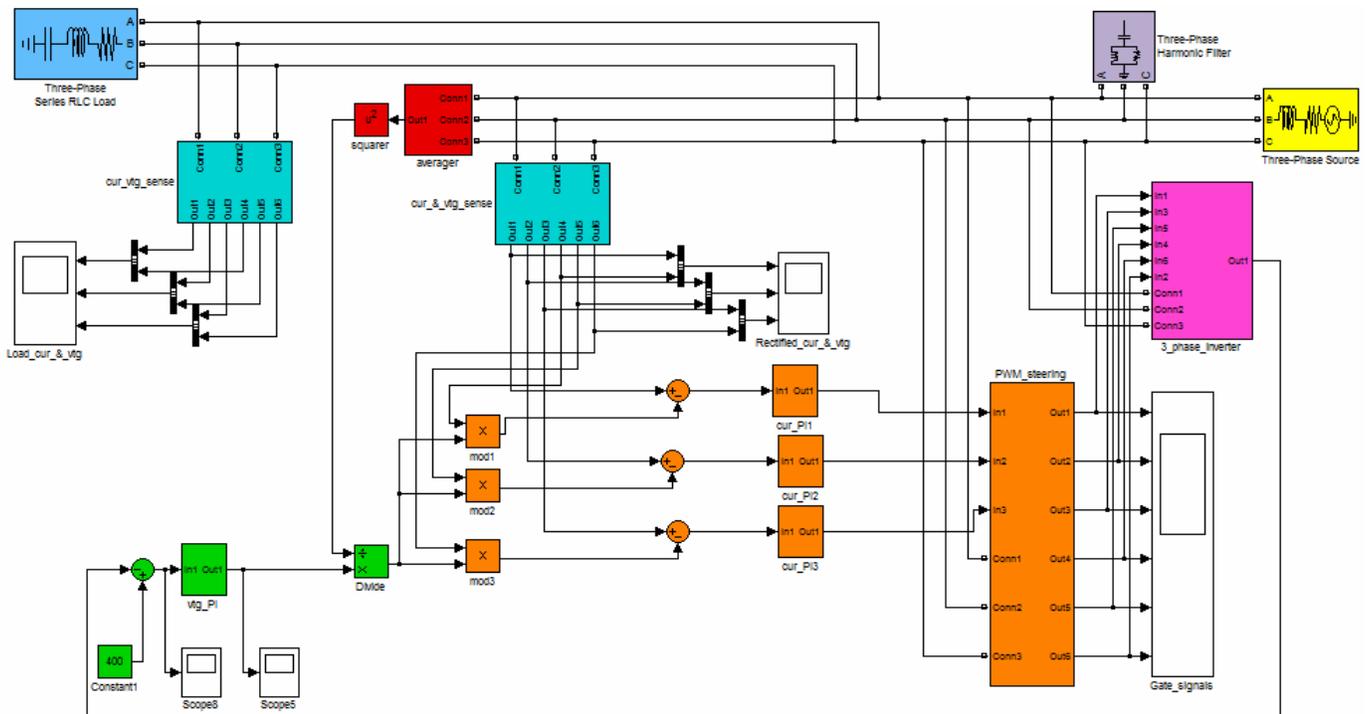


Figure 14: Simulation model of control scheme used for three phase PFC

Outer loop that is voltage control loop generates the reference for the current control loop. According to the error generated in the current error compensator PWM generator

generates pulses with reference to ramp signal. The steering control is used along with ZCD circuit to divide the pulses to drive the IGBT switches.

VI. RESULTS

Fig. 15 shows the power quality drawn without APFC. It can be observed that the poor PF, waveform distortions and harmonics are present. .

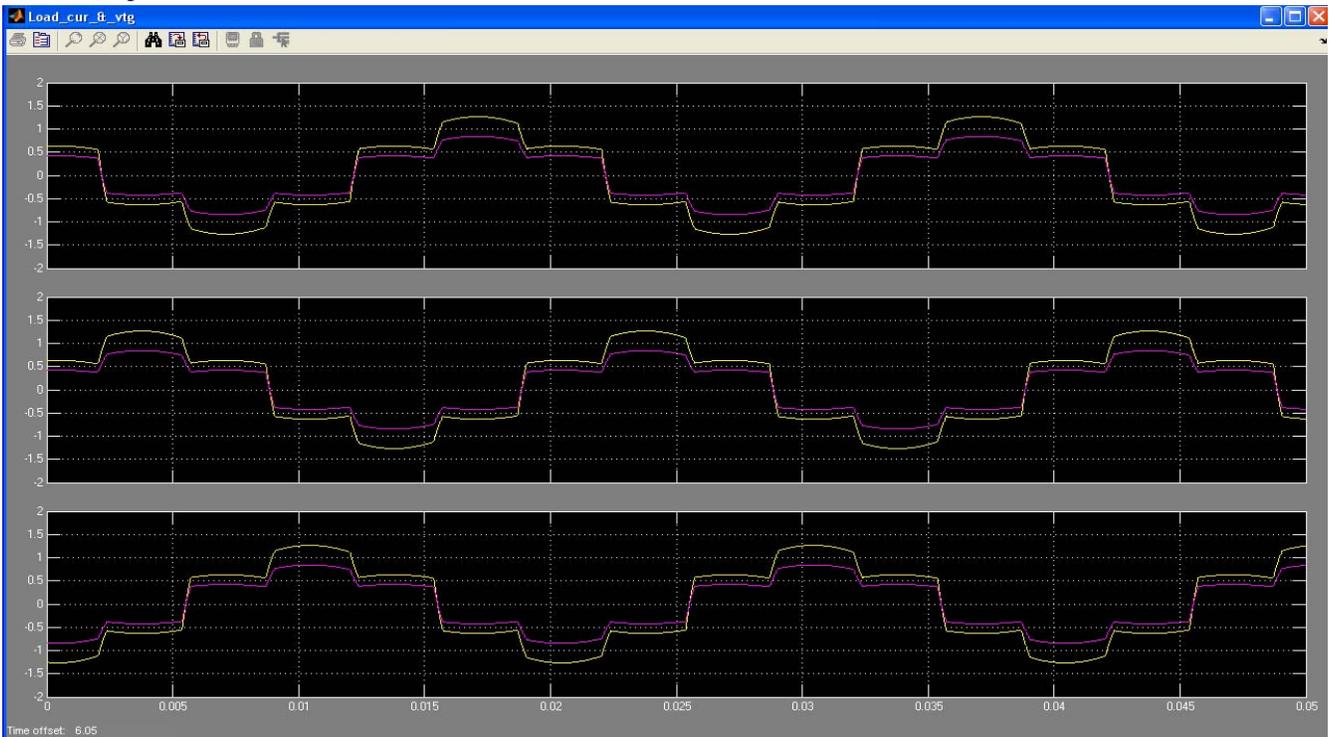


Figure 15: Supply Voltage and Supply Current without Power Factor correction and Harmonic Control

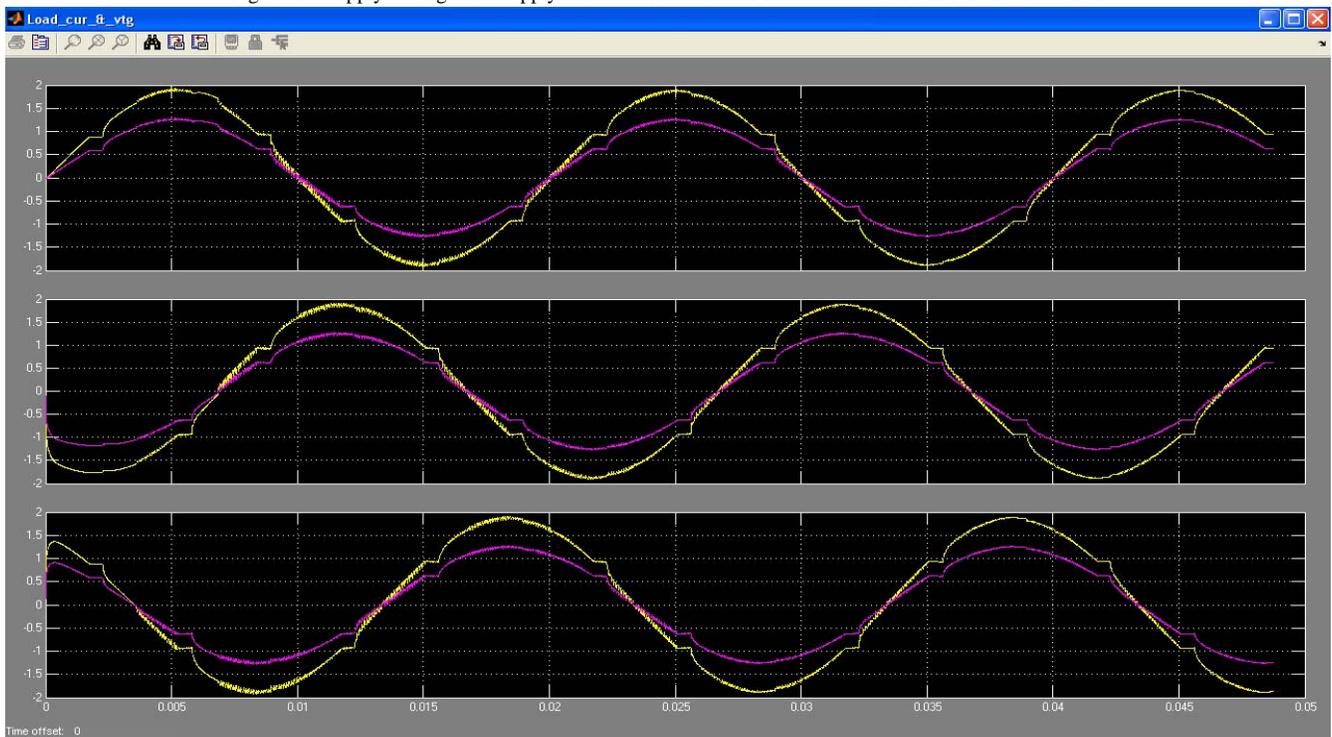


Figure 16: Supply Voltage and Supply Current with Power Factor correction and Harmonic Control

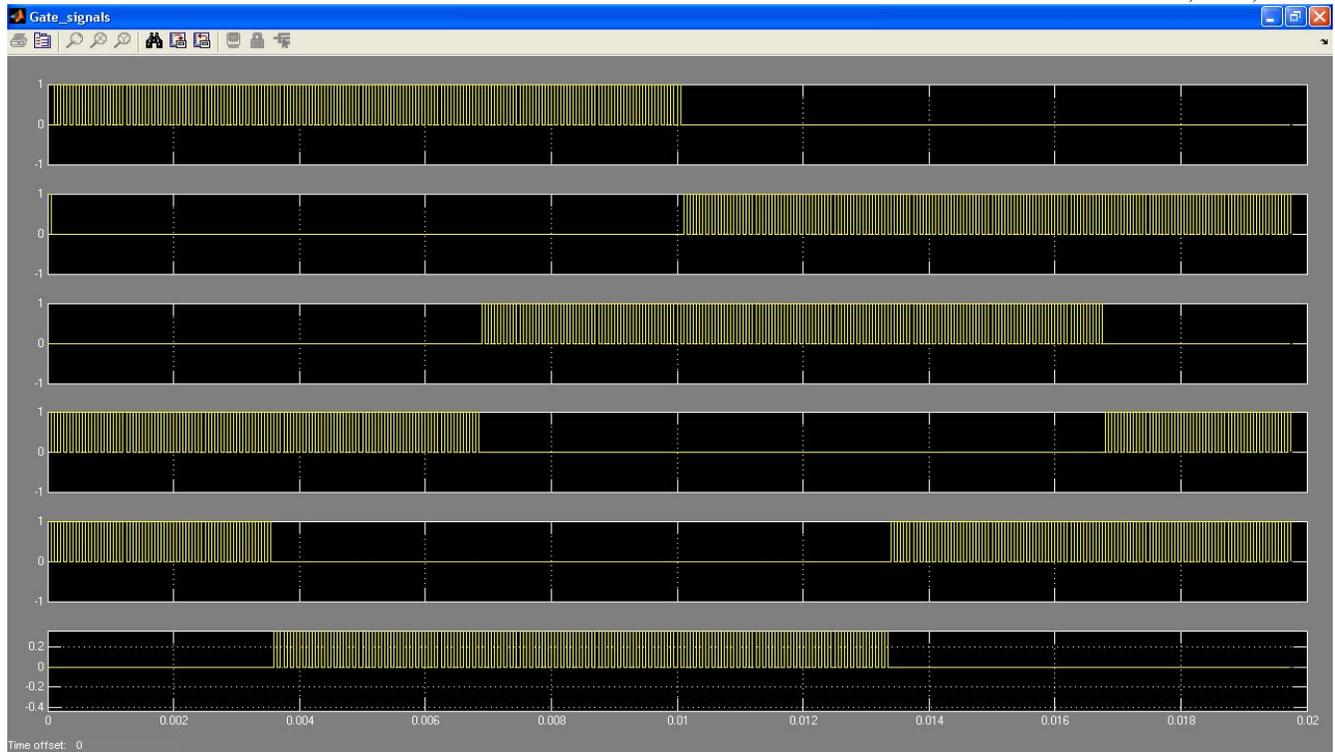


Figure 17: Gate signals for 3 phase inverter or PWM output from control scheme

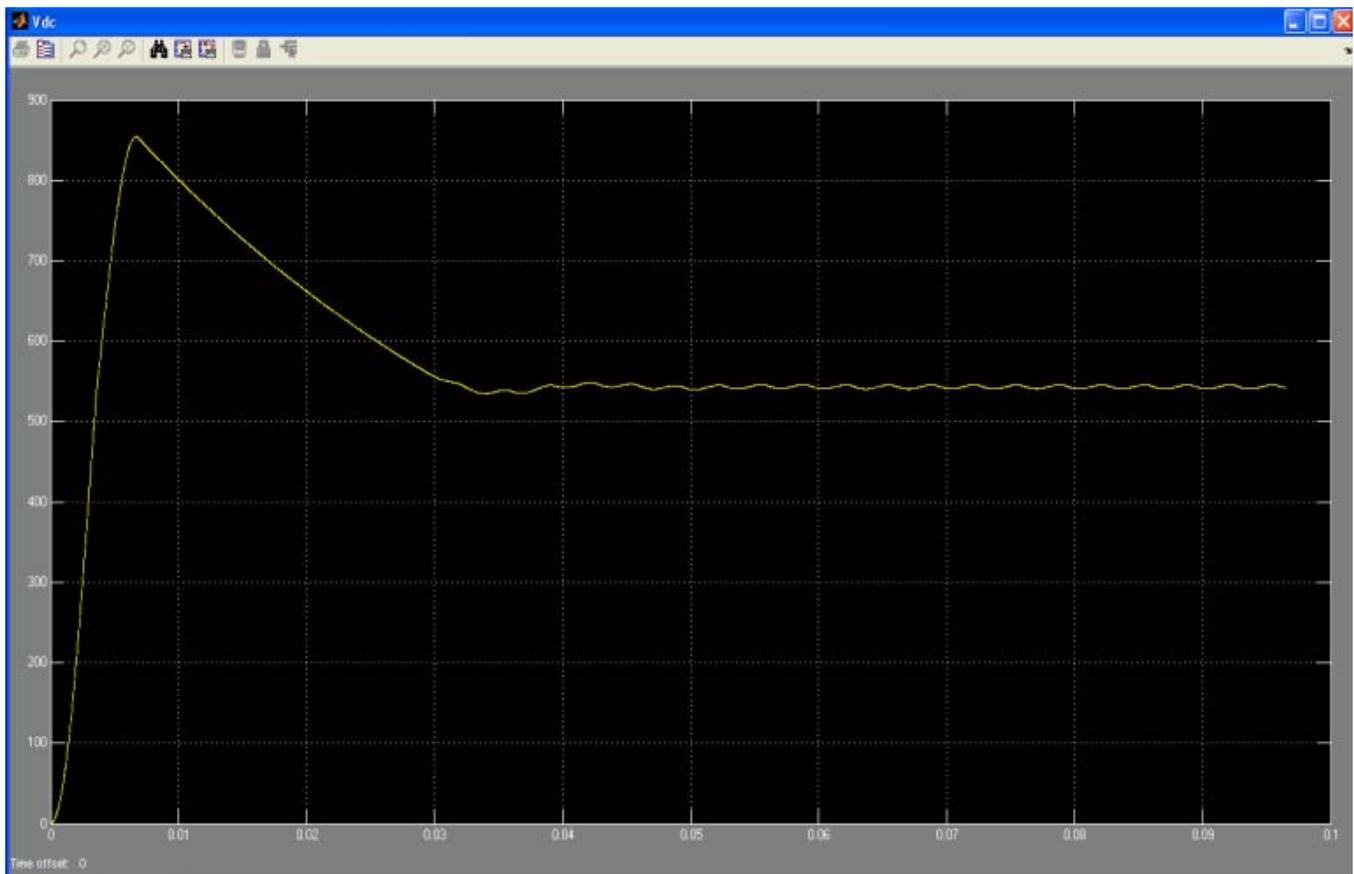


Figure 18: DC Bus Voltage (Capacitor voltage Vdc)

Fig. 16 shows the power quality drawn with APFC. The improvement can be observed in quality of power drawn in terms of PF, waveform distortion and harmonics. Fig. 17 shows the PWM driving signals generated as a result of the current drawn by the load connected.

The resulting power factor of 0.999 and Total Harmonic Distortion (THD) of 3.81%, can be achieved to the 50th line frequency harmonic at nominal line and full load [9].

VII. CONCLUSIONS

Above result analysis shows that the active power filter using switching regulator is a high performance power electronics converter and can operate in different modes: harmonics elimination, power factor correction, voltage regulation and load unbalance compensation. It improves the simplicity and accuracy of control for the active power filter. Different application may lead to use of different switching regulator topology where they add some advantages.

VIII. REFERENCES

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